

信号分析与处理-作业4

1. 已知有限长序列 $x(n)$, $x(n) = \begin{cases} 1 & n=0 \\ 2 & n=1 \\ -1 & n=2 \\ 3 & n=3 \end{cases}$ 又知 $\text{DFT}[x(n)] = X(k) = \sum_{n=0}^3 x(n) W_4^{nk}$, $k=0,1,2,3$

令 $W_4 = e^{-j\frac{2\pi}{4}} = e^{-j\frac{\pi}{2}}$, $W_4^2 = e^{-j\pi} = -1$, $W_4^4 = 1$, $W_4^3 = -j$; $W_4^{r+4} = W_4^r$

$$\begin{pmatrix} X(0) \\ X(1) \\ X(2) \\ X(3) \end{pmatrix} = \begin{pmatrix} W_4^0 & W_4^0 & W_4^0 & W_4^0 \\ W_4^0 & W_4^1 & W_4^2 & W_4^3 \\ W_4^0 & W_4^2 & W_4^4 & W_4^6 \\ W_4^0 & W_4^3 & W_4^6 & W_4^9 \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ x(3) \end{pmatrix} = \begin{pmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{pmatrix} \begin{pmatrix} 1 \\ 2 \\ -1 \\ 3 \end{pmatrix} = \begin{pmatrix} 5 \\ 2+j \\ -5 \\ 2-j \end{pmatrix}$$

再求 $\text{IDFT}[X(k)] = x(n)$, $x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{j\frac{2\pi}{N}nk}$, $N=4, n=0,1,2,3$

$$\begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ x(3) \end{pmatrix} = \frac{1}{4} \begin{pmatrix} W_4^0 & W_4^0 & W_4^0 & W_4^0 \\ W_4^0 & W_4^{-1} & W_4^{-2} & W_4^{-3} \\ W_4^0 & W_4^{-2} & W_4^{-4} & W_4^{-6} \\ W_4^0 & W_4^{-3} & W_4^{-6} & W_4^{-9} \end{pmatrix} \begin{pmatrix} X(0) \\ X(1) \\ X(2) \\ X(3) \end{pmatrix} = \frac{1}{4} \begin{pmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{pmatrix} \begin{pmatrix} 5 \\ 2+j \\ -5 \\ 2-j \end{pmatrix} = \begin{pmatrix} 1 \\ 2 \\ -1 \\ 3 \end{pmatrix}$$

经验证, 结果正确.

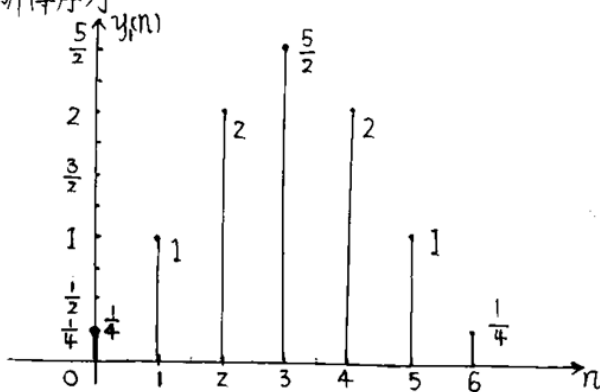
2. 由题可知 $x(n) = \begin{cases} \frac{1}{2} & n=0 \\ 1 & n=1 \\ 1 & n=2 \\ \frac{1}{2} & n=3 \end{cases}$, $N=4$

(1) 求 $x(n)$ 与 $x(n)$ 的线性卷积, $y_1(n) = \sum_{m=-\infty}^{+\infty} x(m) \cdot x(n-m)$

得 $y_1(n) = x(0) \cdot x(n) + x(1) \cdot x(n-1) + x(2) \cdot x(n-2) + x(3) \cdot x(n-3)$

根据已给的 $x(n)$, 可得 $y_1(n) = \frac{1}{4} \delta(n) + 1 \cdot \delta(n-1) + 2 \cdot \delta(n-2) + \frac{5}{2} \delta(n-3) + 2 \delta(n-4) + 1 \delta(n-5) + \frac{1}{4} \delta(n-6)$

可画出所得序列



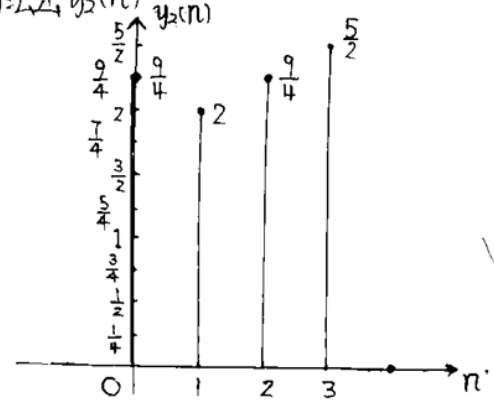
(2) 求 $x(n)$ 与 $x(n)$ 的4点圆周卷积, 圆周卷积

$$y_2(n) = x(n) \otimes x(n) = \sum_{m=0}^3 x(m) \cdot x((n-m)_4) R_4(n)$$

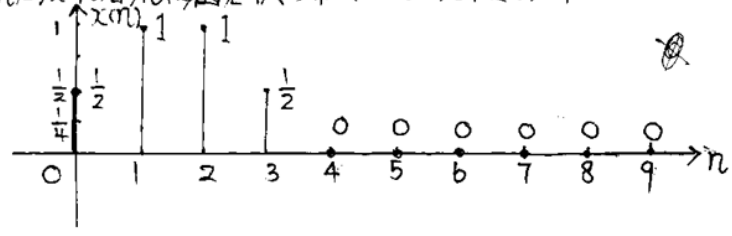
4) 求 \$y_2(n)\$

即 \$y_2(n) = x(0) \cdot x(n) \cdot R_4(n) + x(1) \cdot x(n-1) \cdot R_4(n) + x(2) \cdot x(n-2) \cdot R_4(n) + x(3) \cdot x(n-3) \cdot R_4(n)\$
 可得 \$y_2(n) = \frac{9}{4} \delta(n) + 2\delta(n-1) + \frac{9}{4} \delta(n-2) + \frac{5}{2} \delta(n-3)\$

至此可画出 \$y_2(n)\$



(3) 求 \$x(n)\$ 与 \$x(n)\$ 的 10 点圆卷积，需将 \$x(n)\$ 补 6 个零，如下

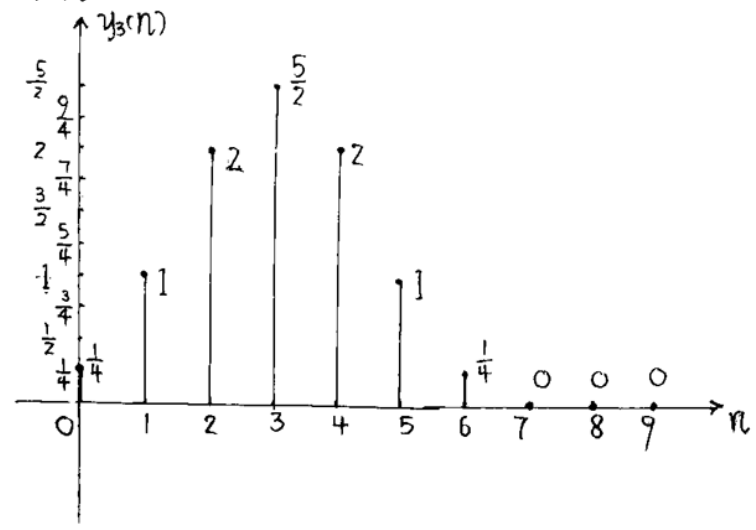


此时 \$y_3(n) = x(n) \otimes x(n) = \sum_{m=0}^9 x(m) \cdot x(n-m) \cdot R_{10}(n)\$

由于 \$x(4) = x(5) = x(6) = x(7) = x(8) = x(9) = 0\$

即 \$y_3(n) = x(0) \cdot x(n) \cdot R_{10}(n) + x(1) \cdot x(n-1) \cdot R_{10}(n) + x(2) \cdot x(n-2) \cdot R_{10}(n) + x(3) \cdot x(n-3) \cdot R_{10}(n)\$
 \$= \frac{1}{4} \delta(n) + 1 \cdot \delta(n-1) + 2 \cdot \delta(n-2) + \frac{5}{2} \cdot \delta(n-3) + 2 \cdot \delta(n-4) + 1 \cdot \delta(n-5) + \frac{1}{4} \delta(n-6) + 0 \cdot \delta(n-7) + 0 \cdot \delta(n-8) + 0 \cdot \delta(n-9)\$

由此可画出 \$y_3(n)\$

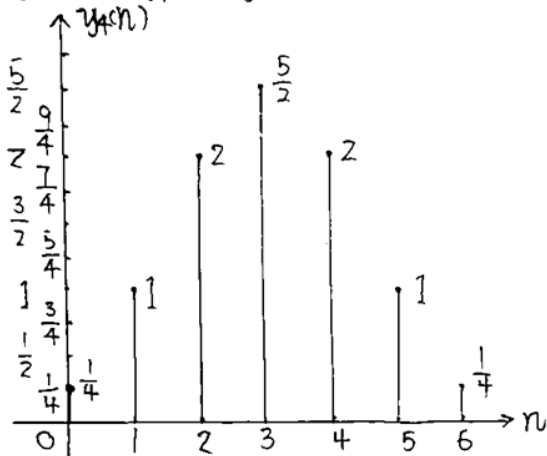


要求 $x(n)$ 与 $x(n)$ 的圆周卷积和线性卷积相同, 由题 $x(n)$ 圆周卷积需补零, 其补零个数应为有效移位数 $4-1=3$, 故 $L_{\min}=4+4-1=7$, 其最小值为 7.

$$x(n) \circledast x(n) = y_4(n) = \sum_{m=0}^6 x(m)x(n-m) R_7(n), \text{ 且 } x(4)=x(5)=x(6)=0$$

$$\begin{aligned} \text{即 } y_4(n) &= x(0) \cdot x(n) R_7(n) + x(1)x(n-1) R_7(n) + x(2)x(n-2) R_7(n) + x(3)x(n-3) R_7(n) \\ &= \frac{1}{4} \delta(n) + 1 \cdot \delta(n-1) + 2 \cdot \delta(n-2) + \frac{5}{2} \delta(n-3) + 2 \delta(n-4) + 1 \cdot \delta(n-5) + \frac{1}{4} \delta(n-6) \end{aligned}$$

由此可绘出 $y_4(n)$, 可知 $y_4(n)$ 与 $y_1(n)$ 一致, L 最小值应为 6.



3.

(1) 已知 $y'(t) + 3y(t) = 2 \cdot x(t)$, 可设 $y(t) \xrightarrow{F} Y(\omega)$, $x(t) \xrightarrow{F} X(\omega)$

将上述方程化为 $j\omega Y(\omega) + 3Y(\omega) = 2j\omega X(\omega)$, 可得 $H(\omega) = \frac{Y(\omega)}{X(\omega)} = \frac{2j\omega}{j\omega + 3} = 2 - 6 \cdot \frac{1}{j\omega + 3}$

则系统冲激响应的傅里叶变换为 $H(\omega)$

$$h(t) = F^{-1}(H(\omega)), \text{ 又知 } F(\delta(t)) = 1, F(e^{-3t} \cdot u(t)) = \frac{1}{j\omega + 3}$$

$$\text{即 } h(t) = 2\delta(t) - 6e^{-3t} \cdot u(t)$$

单位阶跃信号 $u(t)$ 的傅里叶变换为 $\pi\delta(\omega) + \frac{1}{j\omega}$

$$\text{单位阶跃响应 } c(t) \text{ 的傅里叶变换为 } C(\omega) = H(\omega) \cdot (\pi\delta(\omega) + \frac{1}{j\omega}) = \frac{2j\omega}{j\omega + 3} (\pi\delta(\omega) + \frac{1}{j\omega})$$

即 $C(\omega) = \frac{2}{j\omega + 3}$, 由于 $f(\omega) \cdot \delta(\omega) = f(0) \delta(\omega)$, 可知 $H(0) = 0$, 即 $C(\omega)$ 前一项为 0, 只有后一项

$$c(t) = F^{-1}(C(\omega)) = 2e^{-3t} u(t)$$

(2) 已知 $y''(t) + y(t) = x(t) + x(t)$, 同样有 $F(y(t)) = Y(\omega)$, $F(x(t)) = X(\omega)$ 可假设

$$\text{将上述方程化为 } (j\omega)^2 Y(\omega) + Y(\omega) = j\omega X(\omega) + X(\omega), \text{ 得 } H(\omega) = \frac{j\omega + 1}{(j\omega)^2 + j\omega + 1}$$

$$\text{可化 } H(\omega) = \frac{(j\omega + \frac{1}{2}) + \frac{\sqrt{3}}{2}}{(j\omega + \frac{1}{2})^2 + (\frac{\sqrt{3}}{2})^2} = \frac{j\omega + \frac{1}{2}}{(j\omega + \frac{1}{2})^2 + (\frac{\sqrt{3}}{2})^2} + \frac{\sqrt{3}}{3} \cdot \frac{\frac{\sqrt{3}}{2}}{(j\omega + \frac{1}{2})^2 + (\frac{\sqrt{3}}{2})^2}$$

由此 $h(t) = F^{-1}(H(\omega))$, 而 $F^{-1}\left(\frac{j\omega + \frac{1}{2}}{(j\omega + \frac{1}{2})^2 + \frac{3}{4}}\right) = e^{-\frac{1}{2}t} \cdot \cos\frac{\sqrt{3}}{2}t \cdot u(t)$
 $F^{-1}\left(\frac{\frac{\sqrt{3}}{2}}{(j\omega + \frac{1}{2})^2 + \frac{3}{4}}\right) = e^{-\frac{1}{2}t} \cdot \sin\frac{\sqrt{3}}{2}t \cdot u(t)$

综上 $h(t) = e^{-\frac{1}{2}t} \cdot \cos\frac{\sqrt{3}}{2}t \cdot u(t) + \frac{\sqrt{3}}{3} \cdot e^{-\frac{1}{2}t} \cdot \sin\frac{\sqrt{3}}{2}t \cdot u(t)$.

而单位冲激响应的傅里叶变换 $C(\omega) = H(\omega) \cdot (\pi\delta(\omega) + \frac{1}{j\omega}) = \frac{j\omega + 1}{(j\omega)^2 + j\omega + 1} (\pi\delta(\omega) + \frac{1}{j\omega})$

同理 $\frac{j\omega + 1}{(j\omega)^2 + j\omega + 1} \cdot \pi\delta(\omega) = \pi\delta(\omega)$

即 $C(\omega) = \pi\delta(\omega) + \frac{j\omega + 1}{(j\omega)((j\omega)^2 + j\omega + 1)} = \pi\delta(\omega) + \frac{1}{j\omega} + \frac{-(j\omega) - 1}{(j\omega)^2 + j\omega + 1} + \frac{1}{(j\omega)^2 + j\omega + 1}$

前两项 $\pi\delta(\omega) + \frac{1}{j\omega} = F(u(t))$, 第三项 $= -\frac{j\omega + \frac{1}{2}}{(j\omega + \frac{1}{2})^2 + (\frac{\sqrt{3}}{2})^2} - \frac{\frac{\sqrt{3}}{2}}{(j\omega + \frac{1}{2})^2 + (\frac{\sqrt{3}}{2})^2}$
 $= F(-e^{-\frac{1}{2}t} \cdot \cos\frac{\sqrt{3}}{2}t \cdot u(t) - \frac{\sqrt{3}}{3} \cdot e^{-\frac{1}{2}t} \cdot \sin\frac{\sqrt{3}}{2}t \cdot u(t))$

即 $C(t) = u(t) - e^{-\frac{1}{2}t} \cdot \cos\frac{\sqrt{3}}{2}t \cdot u(t) - \frac{\sqrt{3}}{3} e^{-\frac{1}{2}t} \cdot \sin\frac{\sqrt{3}}{2}t \cdot u(t)$

第四项 $= \frac{1}{(j\omega)^2 + j\omega + 1} = \frac{2\sqrt{3}}{3} \frac{\frac{\sqrt{3}}{2}}{(j\omega + \frac{1}{2})^2 + (\frac{\sqrt{3}}{2})^2} = F\left(\frac{2\sqrt{3}}{3} \cdot e^{-\frac{1}{2}t} \cdot \sin\frac{\sqrt{3}}{2}t \cdot u(t)\right)$

故 $C(t) = u(t) - e^{-\frac{1}{2}t} \cdot \cos\frac{\sqrt{3}}{2}t \cdot u(t) + \frac{\sqrt{3}}{3} \cdot e^{-\frac{1}{2}t} \cdot \sin\frac{\sqrt{3}}{2}t \cdot u(t)$

(3) 由题已知 $y'(t) + 2y(t) = x''(t) + 3x'(t) + 3x(t)$, 假设 $F(y(t)) = Y(\omega)$, $F(x(t)) = X(\omega)$
 上述化为 $j\omega \cdot Y(\omega) + 2Y(\omega) = (j\omega)^2 X(\omega) + 3j\omega X(\omega) + 3X(\omega)$, 得 $H(\omega) = \frac{Y(\omega)}{X(\omega)} = (j\omega + 1) + \frac{1}{j\omega + 2}$

$h(t) = F^{-1}(H(\omega)) = \delta(t) + \delta(t) + e^{-2t} \cdot u(t)$

即 $C(\omega) = H(\omega) \cdot (\pi\delta(\omega) + \frac{1}{j\omega}) = \pi \cdot H(0) \cdot \delta(\omega) + \frac{1}{j\omega} H(\omega) = \frac{3}{2} \pi\delta(\omega) + 1 + \frac{1}{j\omega} \frac{3}{2} + \frac{1}{j\omega + 2} \cdot (-\frac{1}{2})$

$C(t) = F^{-1}(C(\omega)) = \frac{3}{4} + \delta(t) + \frac{3}{4} \text{sgn}(t) - \frac{1}{2} e^{-2t} u(t)$

$= \frac{3}{2} u(t) + \delta(t) - \frac{1}{2} e^{-2t} u(t)$

4. 已知 $H(\omega) = \frac{1}{j\omega + 1}$, 激励信号 $e(t) = \sin t + \sin 3t$

由此 $E(\omega) = F(e(t)) = j\pi (\delta(\omega + 1) - \delta(\omega - 1) + \delta(\omega + 3) - \delta(\omega - 3))$

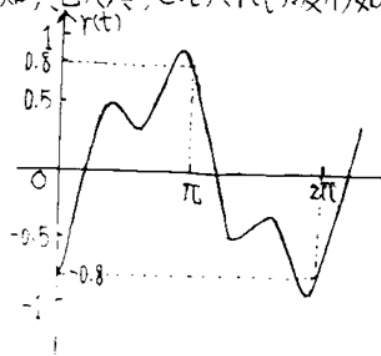
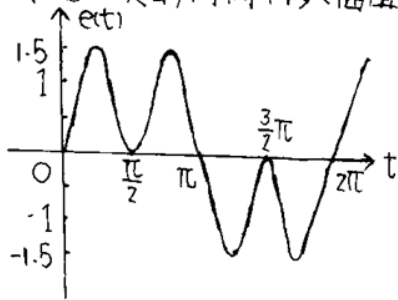
而 $R(\omega) = F(r(t)) = E(\omega) \cdot H(\omega) = \frac{j\pi}{j\omega + 1} (\delta(\omega + 1) - \delta(\omega - 1) + \delta(\omega + 3) - \delta(\omega - 3))$

$= \frac{j\pi}{1-j} \delta(\omega + 1) - \frac{j\pi}{1+j} \delta(\omega - 1) + \frac{j\pi}{1+3j} \delta(\omega + 3) - \frac{j\pi}{1-3j} \delta(\omega - 3)$

解法

$$\begin{aligned}
 r(t) = F^{-1}(F(\omega)) &= \frac{j}{2(1-j)} e^{-jt} - \frac{j}{2(1+j)} e^{jt} + \frac{j}{2(1-3j)} e^{-j3t} - \frac{j}{2(1+3j)} e^{j3t} \\
 &= \frac{j-1}{4} e^{-jt} + \frac{-j-1}{4} e^{jt} + \frac{j-3}{20} e^{-j3t} + \frac{j-3}{20} e^{j3t} \\
 &= -\frac{1}{4}(e^{jt} + e^{-jt}) + \frac{j}{4}(e^{jt} - e^{-jt}) - \frac{3}{20}(e^{j3t} + e^{-j3t}) + \frac{j}{20}(e^{j3t} - e^{-j3t}) \\
 &= -\frac{1}{2} \cos t + \frac{1}{2} \sin t - \frac{3}{10} \cos 3t + \frac{1}{10} \sin 3t \\
 &= \frac{\sqrt{2}}{2} \sin(t-45^\circ) + \frac{\sqrt{10}}{10} \sin(3t-71.565^\circ)
 \end{aligned}$$

由于前一项、后一项的时间平移、幅值变化不一，可知其已失真， $e(t)$ 、 $r(t)$ 波形如下：



由此传输发生失真， $r(t)$ 不仅幅值失真，还有相位失真。

5. 已知 $H(\omega) = \begin{cases} 1 & |\omega| < \frac{2\pi}{T} \\ 0 & |\omega| > \frac{2\pi}{T} \end{cases}$ ，且 $X(\omega) = T \text{Sa}(\frac{\omega T}{2})$ 作为激励

由此 $h(t) = F^{-1}(H(\omega)) = \frac{2}{T} \text{Sa}(\frac{2\pi}{T}t)$ ，而 $x(t) = \begin{cases} 1 & |t| < \frac{T}{2} \\ 0 & |t| > \frac{T}{2} \end{cases} = u(t + \frac{T}{2}) - u(t - \frac{T}{2})$

$$\begin{aligned}
 \text{故 } y(t) = h(t) * x(t) &= \frac{2}{T} \text{Sa}(\frac{2\pi}{T}t) * (u(t + \frac{T}{2}) - u(t - \frac{T}{2})) \\
 &= \int_{-\infty}^{+\infty} \frac{2}{T} \text{Sa}(\frac{2\pi}{T}\lambda) \cdot u(t + \frac{T}{2} - \lambda) d\lambda - \int_{-\infty}^{+\infty} \frac{2}{T} \text{Sa}(\frac{2\pi}{T}\lambda) \cdot u(t - \frac{T}{2} - \lambda) d\lambda \\
 &= \int_{-\infty}^{t + \frac{T}{2}} \frac{2}{T} \text{Sa}(\frac{2\pi}{T}\lambda) d\lambda - \int_{-\infty}^{t - \frac{T}{2}} \frac{2}{T} \text{Sa}(\frac{2\pi}{T}\lambda) d\lambda \\
 &\stackrel{\text{令 } x = \frac{2\pi}{T}\lambda}{=} \frac{1}{\pi} \int_{-\infty}^{\frac{2\pi}{T}(t + \frac{T}{2})} \text{Sa}(x) dx - \frac{1}{\pi} \int_{-\infty}^{\frac{2\pi}{T}(t - \frac{T}{2})} \text{Sa}(x) dx \\
 &= \frac{1}{\pi} \left(\text{Si}(\frac{2\pi}{T}(t + \frac{T}{2})) - \text{Si}(\frac{2\pi}{T}(t - \frac{T}{2})) \right)
 \end{aligned}$$

6. 由频域指标，可有：假设 $\omega_c = \omega_1 = \omega_p = 1000 \text{ rad/s}$ ， $\alpha_p = 3 \text{ dB}$ 。

$\omega_s = \omega_2 = 5000 \text{ rad/s}$ ， $\alpha_s = 20 \text{ dB}$

见背面

求巴特沃斯滤波器阶数

$$n = \frac{\lg \sqrt{10^{0.1n_s} - 1}}{\lg \frac{\omega_s}{\omega_c}} = 1.428. \text{ 故取 } n=2, \text{ 查巴特沃斯多项式表后}$$

$$\text{得 } H(\bar{s}) = \frac{1}{\bar{s}^2 + \sqrt{2}\bar{s} + 1} = \frac{\omega_c^2}{\bar{s}^2 + \sqrt{2}\omega_c\bar{s} + \omega_c^2} = \frac{10^6}{\bar{s}^2 + \sqrt{2} \times 10^3\bar{s} + 10^6}$$

通过反归一化处理, 令 $s = \bar{s}\omega_c$ 得

$$H(s) = \frac{\omega_c^2}{s^2 + \sqrt{2}\omega_c s + \omega_c^2} = \frac{10^6}{s^2 + \sqrt{2} \times 10^3 s + 10^6}$$