

$$\lg(AB) = \lg A + \lg B$$

2024 年 12 月 13 日

1 考虑单位反馈系统，其开环传递函数如下，

$$G(s) = \frac{\omega_n^2}{s(s+2\zeta\omega_n)}$$

当取 $r(t) = 2\sin t$ 时，系统的稳态输出

$$\text{输入正弦信号} \rightarrow \text{频域分析} \quad c_s(t) = 2\sin(t - 45^\circ)$$

试确定系统参数 ω_n, ζ 。

解：系统的闭环传递函数为 $\bar{G}(s) = \frac{Y(s)}{X(s)} = \frac{G(s)}{1+G(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$ 为标准的振荡环节

输入 $r(t) = 2\sin t$ ，输出 $c_s(t) = 2\sin(t - 45^\circ)$ ，有 $W = 1 \text{ rad/s}$, $|\bar{G}(j\omega)| = 1$, $\angle \bar{G}(j\omega) = -45^\circ$

$$\bar{G}(j\omega) = \frac{\omega_n^2}{(\omega_n^2 - \omega^2)^2 + (2\zeta\omega_n\omega)^2} \Rightarrow |\bar{G}(j\omega)| = \frac{\omega_n^2}{\sqrt{(\omega_n^2 - \omega^2)^2 + (2\zeta\omega_n\omega)^2}}, \angle \bar{G}(j\omega) = -\arctan \frac{2\zeta\omega_n\omega}{\omega_n^2 - \omega^2}$$

将 $\omega = 1 \text{ rad/s}$ 时 $|\bar{G}(j\omega)| = 1$, $\angle \bar{G}(j\omega) = -45^\circ$ 代入，有

$$\begin{cases} \omega_n^2 = \sqrt{(\omega_n^2 - 1)^2 + (2\zeta\omega_n)^2} \\ 2\zeta\omega_n = \omega_n^2 - 1 \end{cases} \Rightarrow \begin{cases} (4\zeta^2 - 2)\omega_n^2 + 1 = 0 \\ \omega_n^2 - 2\zeta\omega_n - 1 = 0 \end{cases} \Rightarrow \begin{cases} \omega_n^4 - 4\omega_n^2 + 2 = 0 \\ \zeta = \frac{\omega_n^2 - 1}{2\omega_n} \end{cases}$$

$$\text{解得 } \begin{cases} \omega_n = \sqrt{2 + \sqrt{2}} \approx 1.848 \\ \zeta = \frac{1 + \sqrt{2}}{2\sqrt{2 + \sqrt{2}}} \approx 0.653 \end{cases} \text{ 或 } \begin{cases} \omega_n = \sqrt{2 - \sqrt{2}} \approx 0.765 \\ \zeta = \frac{1 - \sqrt{2}}{2\sqrt{2 - \sqrt{2}}} < 0 \end{cases} \text{ 舍去}$$

$$\text{综上, } \omega_n = 1.848, \zeta \approx 0.653$$

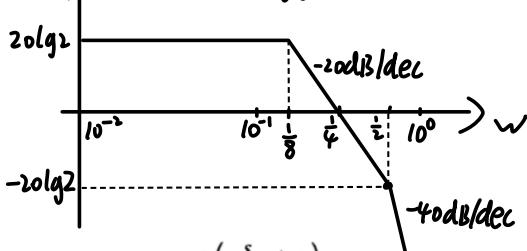
2 绘制下列传递函数的对数幅频渐近特性曲线 一定注意 $10^0 = 1$

$$(1) G(s) = \frac{2}{(2s+1)(8s+1)}; \quad \text{Bode 图}$$

解：已是标准型，转折频率 $\begin{cases} \omega_1 = \frac{1}{8} & -20 \text{ dB/dec} \\ \omega_2 = \frac{1}{2} & -20 \text{ dB/dec} \end{cases}$

基准点： $w=1$ 时 $L(1) = 20\lg k = 20\lg 2 \approx 6.02$ ，余率 $= 0$

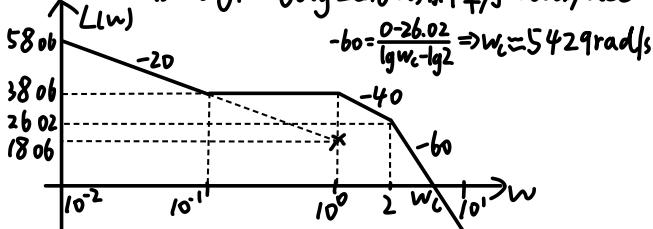
过零点 $k = -20 = \frac{0 - 20\lg 2}{\lg \omega_c - \lg \frac{1}{8}} \Rightarrow$ 剪切频率 $\omega_c = \frac{1}{4} \text{ rad/s}$



$$(3) G(s) = \frac{8\left(\frac{s}{0.1} + 1\right)}{s(s^2 + s + 1)\left(\frac{s}{2} + 1\right)};$$

解：转折频率 $\begin{cases} 0.1 & +20 \text{ dB/dec} \\ 1 & -40 \text{ dB/dec} \\ 2 & -20 \text{ dB/dec} \end{cases}$

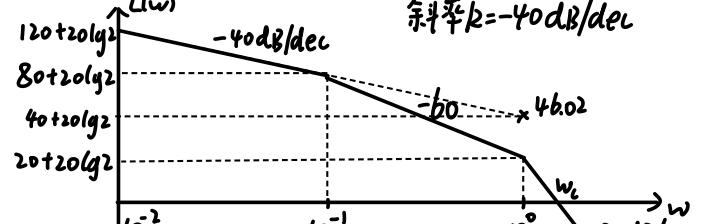
基准点 $L(1) = 20\lg k = 60\lg 2 \approx 18.06$ ，余率 -20 dB/dec



$$(2) G(s) = \frac{200}{s^2(s+1)(10s+1)};$$

已是标准型，转折频率 $\begin{cases} \frac{1}{10} & -20 \text{ dB/dec} \\ 1 & -20 \text{ dB/dec} \end{cases}$

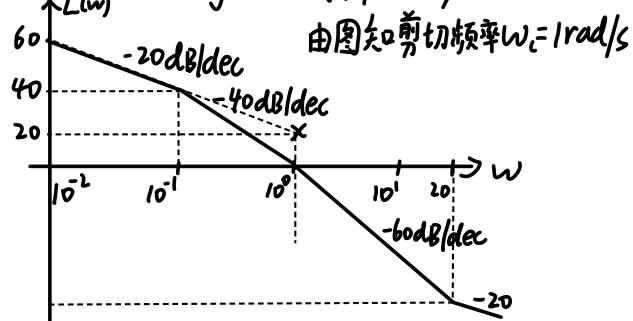
基准点 $L(1) = 20\lg k = 20\lg 200 = 40 + 20\lg 2 \approx 46.02$ ，余率 $k = -40 \text{ dB/dec}$



$$\text{过零点: } k = -80 = \frac{0 - (20 + 20\lg 2)}{\lg \omega_c - \lg 1} \Rightarrow \omega_c = 2.115 \text{ rad/s}$$

$$(4) G(s) = \frac{10\left(\frac{s^2}{400} + \frac{s}{10} + 1\right)}{s(s+1)\left(\frac{s}{0.1} + 1\right)} \quad \text{转折频率} \begin{cases} 0.1 & -20 \text{ dB/dec} \\ 1 & -20 \text{ dB/dec} \\ 20 & 40 \text{ dB/dec} \end{cases}$$

基准点 $L(1) = 20\lg k = 20$ ，余率 -20 dB/dec



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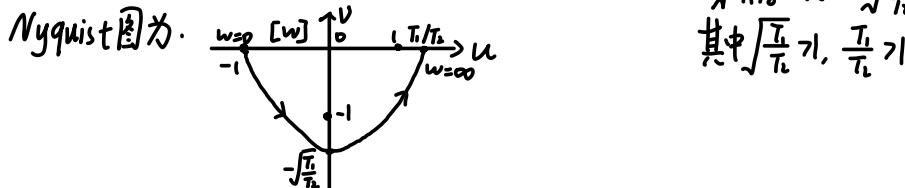
5.1 一阶环节的传递函数为

$$G(s) = \frac{T_1 s + 1}{T_2 s - 1} \quad 1 > T_1 > T_2 > 0$$

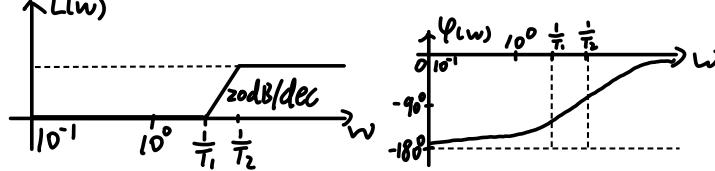
试绘制该环节的 Nyquist 图及 Bode 图。

$$\text{角平分图: } G(s) = \frac{T_1 s + 1}{T_2 s - 1}, \quad 1 > T_1 > T_2 > 0, \quad G(j\omega) = \frac{j\omega T_1 + 1}{j\omega T_2 - 1} = \frac{T_1 T_2 \omega^2 - 1}{T_1^2 \omega^2 + 1} + j \frac{-(T_1 + T_2) \omega}{T_1^2 \omega^2 + 1} = X + jY$$

$$\text{实部 } X = \frac{T_1 T_2 \omega^2 - 1}{T_1^2 \omega^2 + 1}, \quad \text{虚部 } Y = \frac{-(T_1 + T_2) \omega}{T_1^2 \omega^2 + 1}, \quad \text{交点 } \omega=0 \text{ 时 } (-1, 0), \quad \omega = \sqrt{\frac{1}{T_1 T_2}} \text{ 时 } (0, -\sqrt{\frac{T_1}{T_2}}), \quad \omega = \infty \text{ 时 } (\frac{T_1}{T_2}, 0)$$



转折频率 $\left\{ \begin{array}{l} 1 < \frac{1}{T_1} (\text{下}) + 20 \text{dB/dec} \\ 1 < \frac{1}{T_2} (\text{上}) - 20 \text{dB/dec} \end{array} \right.$ 基准线斜率为 0 中频角 $\omega=0$ 时为 -180° $\omega=\infty$ 时为 0°



5.3 设某系统的开环传递函数为

$$G(s)H(s) = \frac{Ke^{-0.1s}}{s(0.1s+1)(s+1)}$$

试通过该系统的频率响应确定剪切频率 $\omega_c = 5 \text{ rad/s}$ 时的开环增益 K 。

$$\text{角平分图: } G(s)H(s) = \frac{Ke^{-0.1j\omega}}{j\omega(0.1j\omega+1)(j\omega+1)} \Rightarrow |G(s)H(s)| = \frac{K}{\omega\sqrt{0.01\omega^2+1}\sqrt{\omega^2+1}}$$

$$\omega_c = 5 \text{ rad/s} \Rightarrow \omega = \omega_c \text{ 时, } |G(s)H(s)| = 1$$

$$\text{代入得 } K = \omega_c \sqrt{0.01\omega_c^2+1} \sqrt{\omega_c^2+1} = 28504$$

5.4 若系统的单位阶跃响应为

$$y(t) = 1 - 1.8e^{-4t} + 0.8e^{-9t} \quad t \geq 0$$

试求取该系统的频率响应 $G(j\omega) = \Phi(s)|_{s=j\omega}$

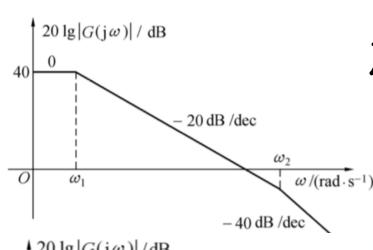
$$\text{角平分图: 系统输入为 } u(t) \xrightarrow{\text{L}} \frac{1}{5}, \text{ 车削出 } y(0)=0, \text{ 零状态, } Y(s) = \mathcal{L}[y(t)] = \frac{1}{5} - 1.8 \frac{1}{s+4} + 0.8 \frac{1}{s+9}$$

$$\text{系统传递函数 } \Phi(s) = \frac{Y(s)}{U(s)} = 1 - \frac{1.8s}{s+4} + \frac{0.8s}{s+9} = \frac{36}{(s+4)(s+9)} = \frac{1}{(\frac{s}{4}+1)(\frac{s}{9}+1)}$$

$$\text{由于 } p_1 = -4, p_2 = -9, \text{ 系统稳定, 系统的频率响应为 } G(j\omega) = \Phi(s)|_{s=j\omega} = \frac{1}{(\frac{j\omega}{4}+1)(\frac{j\omega}{9}+1)}$$

5.5 已知最小相位系统 Bode 图的幅频特性如题 5.5 图所示。试求取该系统的开环传递函

数。



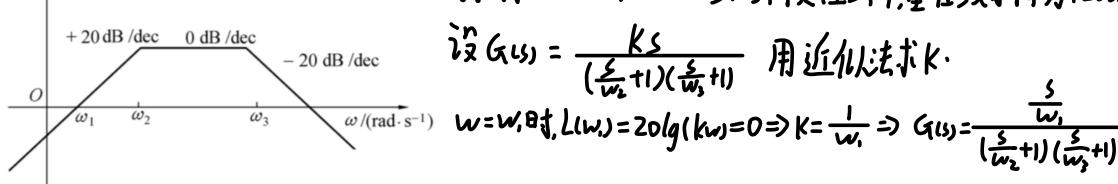
角平分图: 基准线 $L(1) = 20 \lg k = 40 \Rightarrow k = 100$, 斜率为 0, 无积分环节

$\omega = \omega_1, \omega = \omega_2$ 处有两个惯性环节, 又由于是最小相位系统

$$\text{开环传递函数 } G(s) = \frac{100}{(\frac{s}{\omega_1}+1)(\frac{s}{\omega_2}+1)}$$

角平分图: 有 $\omega = \omega_1, \omega = \omega_3$ 两个惯性环节, 基准线斜率为 $+20 \text{dB/dec}$

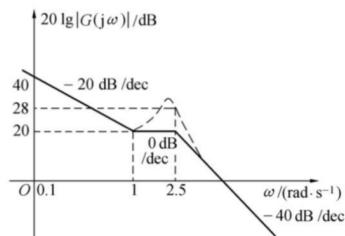
$$\text{设 } G(s) = \frac{ks}{(\frac{s}{\omega_2}+1)(\frac{s}{\omega_3}+1)} \text{ 用近似法求 } k.$$



$$\omega = \omega_1 \text{ 时, } L(\omega_1) = 20 \lg(k\omega_1) = 0 \Rightarrow k = \frac{1}{\omega_1} \Rightarrow G(s) = \frac{\frac{s}{\omega_1}}{(\frac{s}{\omega_2}+1)(\frac{s}{\omega_3}+1)}$$

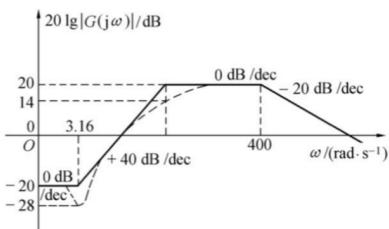
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修正项公式 振荡 $-20\lg(2\xi) \text{ dB}$
二阶微分 $20\lg(2\xi) \text{ dB}$



角频率基准线斜率为 -20 dB/dec , $\omega=1 \text{ rad/s}$, $L(1)=20\lg(k)=20 \Rightarrow k=10$
 $\omega=1$ 处一阶微分环节, $\omega_n=2.5$, 此处 $-20\lg(2\xi)=8 \text{ dB}$

$$\text{传递函数 } G(s) = \frac{10(s+1)}{s[(\frac{s}{2.5})^2 + \frac{0.398s}{2.5} + 1]} \Rightarrow \xi = 0.199$$



-28 为真实曲线的最小值

解: 基准线斜率为 0, $L(1)=20\lg(k)=-20 \Rightarrow k=1$

$\omega=3.16$ 处有一阶微分环节, $\omega_n=31.6$ 处有一振荡环节
(ξ_1) (ξ_2)

$\omega=400$ 处有一惯性环节,

$$G(s) = \frac{(\frac{s}{31.6})^2 + \frac{2\xi_1 s}{31.6} + 1}{10[(\frac{s}{31.6})^2 + \frac{2\xi_1 s}{31.6} + 1](\frac{s}{400} + 1)} \text{ 修正 } \xi_1: 20\lg(2\xi_1) = 8 \Rightarrow \xi_1 = 0.199, \\ \xi_1: 20\lg \frac{1}{2\xi_1 \sqrt{1-\xi_1^2}} = 8 \Rightarrow \xi_1 = 0.203$$

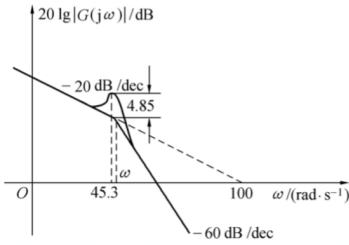
$$G(s) = \frac{(\frac{s}{31.6})^2 + \frac{0.406s}{31.6} + 1}{10[(\frac{s}{31.6})^2 + \frac{1.996s}{31.6} + 1](\frac{s}{400} + 1)} \quad \xi_2: -20\lg(2\xi_2) = -6 \Rightarrow \xi_2 = 0.998$$

角频率基准线斜率为 -20 dB/dec , $L(1)=20\lg k=40 \Rightarrow k=100$

$\omega=\omega_n$ 处有一振荡环节, 参数为 ω_n, ξ ,

$$\text{由 } \begin{cases} \omega_n \sqrt{1-2\xi^2} = 45.3 \\ 20\lg \frac{1}{2\xi \sqrt{1-\xi^2}} = 4.85 \end{cases} \Rightarrow \begin{cases} \omega_n \\ \xi \end{cases} = 0.95 \text{ 估计}, \begin{cases} \omega_n = 50.02 \text{ rad/s} \\ \xi = 0.2999 \end{cases}$$

$$\text{综上, } G(s) = \frac{100}{s[(\frac{s}{50.02})^2 + \frac{0.5998s}{50.02} + 1]}$$



当 $0 < \xi < 0.707$ 时, 对于无零点 $k=1$ 的
标准二阶系统, $\omega_r = \omega_n \sqrt{1-2\xi^2}$

$$M_r = \frac{1}{2\xi \sqrt{1-\xi^2}}$$